

# LINE SEARCH COMPUTATION OF THE BLOCK FACTOR MODEL FOR BLIND MULTI-USER ACCESS IN WIRELESS COMMUNICATIONS



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We present a technique for the computation of the Block Factor Model, which has been introduced in [1] as a powerful **blind receiver** for DS-CDMA signals received on an antenna array, in the context of multi-path propagation with Inter Symbol Interference (ISI). This receiver relies on a **new third-order tensor decomposition** which is computed by an ALS algorithm improved by a Line Search scheme.

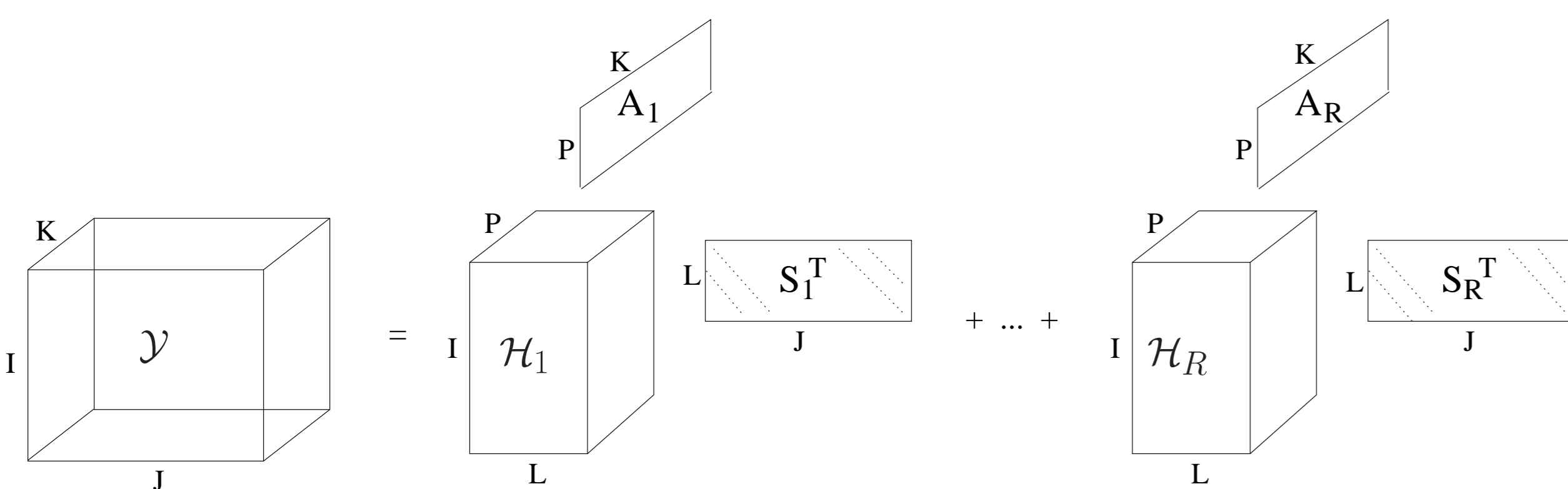
## Communication System

- **Blind Signal Separation: Why?** Estimation of the data relative to each user **without** prior knowledge of the learning sequence.
  - Get higher communication rate.
  - Eavesdropping.
  - Source localisation.
  - Case of learning sequence unavailable or partially received.
- **Parameters and propagation model:**
  - $R$ : Nb of users, transmitting at the same time within the same bandwidth.
  - $I$ : Spreading Factor of CDMA codes.
  - $J$ : Duration of the observation window (in Symbol Periods).
  - $I \times J$  samples collected at the receiver.
  - $K$ : Nb of receiving Antennas.
  - $P$ : Nb of reflected paths per user (Multipath Propagation).
  - $L$ : Nb of interfering symbols (Inter Symbol Interference, ISI).
- **Chip-Rate Sampled Received Signal: Analytic Form**

$$y_{ijk} = \sum_{r=1}^R \sum_{p=1}^P a_k(\theta_{rp}) \sum_{l=1}^L h_{rp}(i + (l-1)I) s_{j-l+1}^{(r)}$$

Labels in diagram: Antenna Response, Channel Effect, Transmitted Symbol, Contribution of R users, Contribution of P paths, Contribution of L ISI.

## Chip-Rate Sampled Received Signal: Algebraic Form



Block Factor Model (BFM)

- The Problem consists of the decomposition of the observation tensor in a sum of  $R$  terms.
- Each Term contains the information related to one particular user (channel, antenna response and symbols).
- The Toeplitz structure of each  $S_r$  is exploited.

## Uniqueness of the Decomposition

- If BFM unique (up to some trivial indeterminacies): separation of the different user signals and estimation of the transmitted sequences are possible.
- Sufficient condition for uniqueness:

$$\min\left(\left\lfloor \frac{J}{L} \right\rfloor, R\right) + \min\left(\left\lfloor \frac{K}{P} \right\rfloor, R\right) + \min\left(\left\lfloor \frac{I}{\max(L, P)} \right\rfloor, R\right) \geq 2R + 2, \quad (1)$$

- Upper bound on the number of users = maximal value of  $R$  that satisfies this equation.  
 Example:  $I = 12, J = 50, K = 4, L = 2, P = 2$ , then  $R = 6$  users can be allowed so more users than antennas is possible.

## Line Search Computation of the Decomposition

- **Objective:** Given only  $\mathcal{Y}$ , estimate  $\mathcal{H}_r, S_r$  and  $\mathbf{A}_r$  for each user.
- Matrix Representation of the unknowns:  $\mathbf{A}$  ( $K \times RP$ ),  $\mathbf{H}$  ( $RLP \times I$ ),  $\mathbf{S}$  ( $J \times RL$ ).
- **Optimization Problem to Solve:** Minimize the cost function

$$\phi_{ALS} = \|\mathcal{Y} - \mathcal{Y}^{(n)}\|^2 = \|\mathcal{Y}^{(JK \times I)} - (\mathbf{S}^{(n)} \circledast \mathbf{A}^{(n)}) \mathbf{H}^{(n)}\|^2, \quad (2)$$

where the superscript  $n$  denotes the estimation at the  $n^{\text{th}}$  iteration.

- **Previous Algorithm:** Alternating Least Squares Algorithm (ALS)  
 Exploit the multilinearity of the model to alternate between conditional least-squares updates of  $\mathbf{A}$ ,  $\mathbf{S}$  and  $\mathbf{H}$ .
- **Improvement scheme:** Iterative Line Search + ALS (ILS + ALS)  
 Idea: Perform Linear Regression to predict the factors a number of iterations ahead before each ALS iteration.

$$\begin{cases} \mathbf{A}^{(new)} = \mathbf{A}^{(n-2)} + \rho(\mathbf{A}^{(n-1)} - \mathbf{A}^{(n-2)}) \\ \mathbf{S}^{(new)} = \mathbf{S}^{(n-2)} + \rho(\mathbf{S}^{(n-1)} - \mathbf{S}^{(n-2)}) \\ \mathbf{H}^{(new)} = \mathbf{H}^{(n-2)} + \rho(\mathbf{H}^{(n-1)} - \mathbf{H}^{(n-2)}) \end{cases}, \quad (3)$$

Then give  $\mathbf{A}^{(new)}, \mathbf{S}^{(new)}$  and  $\mathbf{H}^{(new)}$  instead of  $\mathbf{A}^{(n-1)}, \mathbf{S}^{(n-1)}$  and  $\mathbf{H}^{(n-1)}$  as inputs of ALS.

- **Problem:** Find the optimal value of  $\rho \in \mathbb{C}$  that minimizes

$$\phi_{ILS}^{(n)} = \|\mathbf{S}^{(new)} \circledast \mathbf{A}^{(new)} \mathbf{H}^{(new)} - \mathcal{Y}^{(JK \times I)}\|^2 = \mathbf{u}^H \Delta \mathbf{u}, \quad (4)$$

where  $\mathbf{u} = [\rho^3 \ \rho^2 \ \rho \ 1]^T$  and  $\Delta$  is a  $4 \times 4$  known Hermitian matrix.

- **Solution:** Pose  $\rho = r \cdot e^{i\theta}$  and minimize  $\phi_{ILS}^{(n)}$  alternately w.r.t.  $r$  and  $\theta$ .

**Iterative Line Search Scheme:**

1. Partial minimization of  $\phi_{ILS}^{(n)}$

(i) w.r.t.  $r$ :  $\frac{\delta \phi_{ILS}^{(n)}(r)}{\delta r} = \sum_{p=0}^5 c_p r^p$       (ii) w.r.t.  $\theta$  (pose  $t = \tan(\frac{\theta}{2})$ ):  $\frac{\delta \phi_{ILS}^{(n)}(t)}{\delta t} = \frac{\sum_{p=0}^6 d_p t^p}{(1+t^2)^3}$

2. Repeat steps (i) and (ii) until  $\|\phi_{ILS}^{(n)} - \phi_{ILS}^{(n-1)}\| < \eta$  (e.g.  $\eta = 10^{-1}$ )

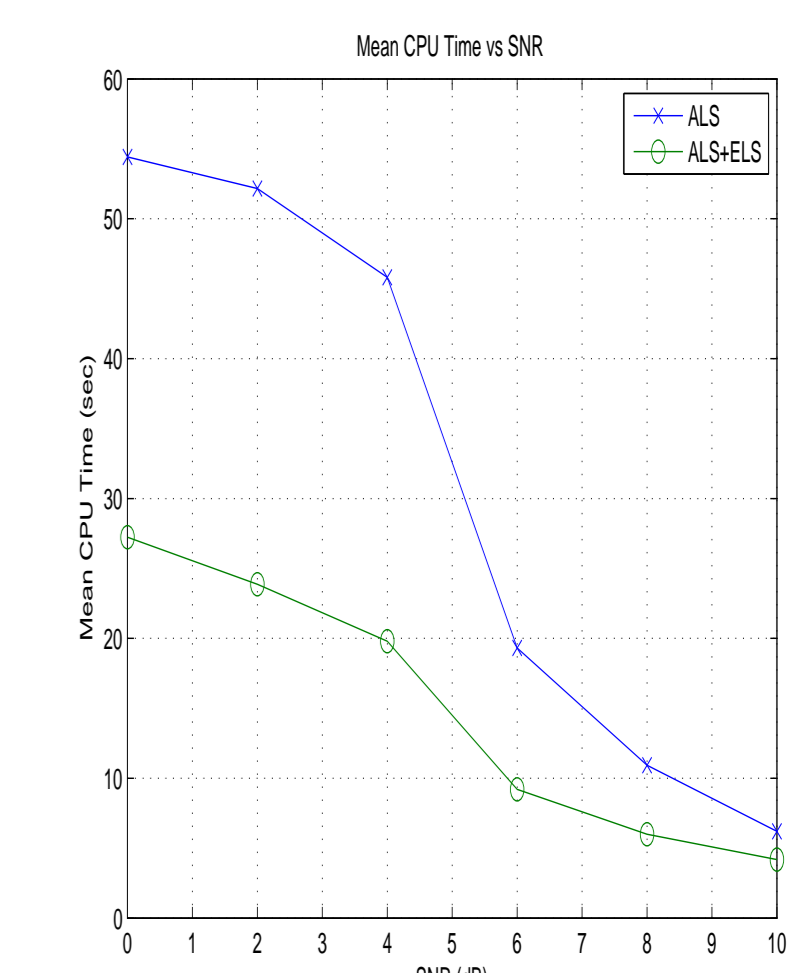
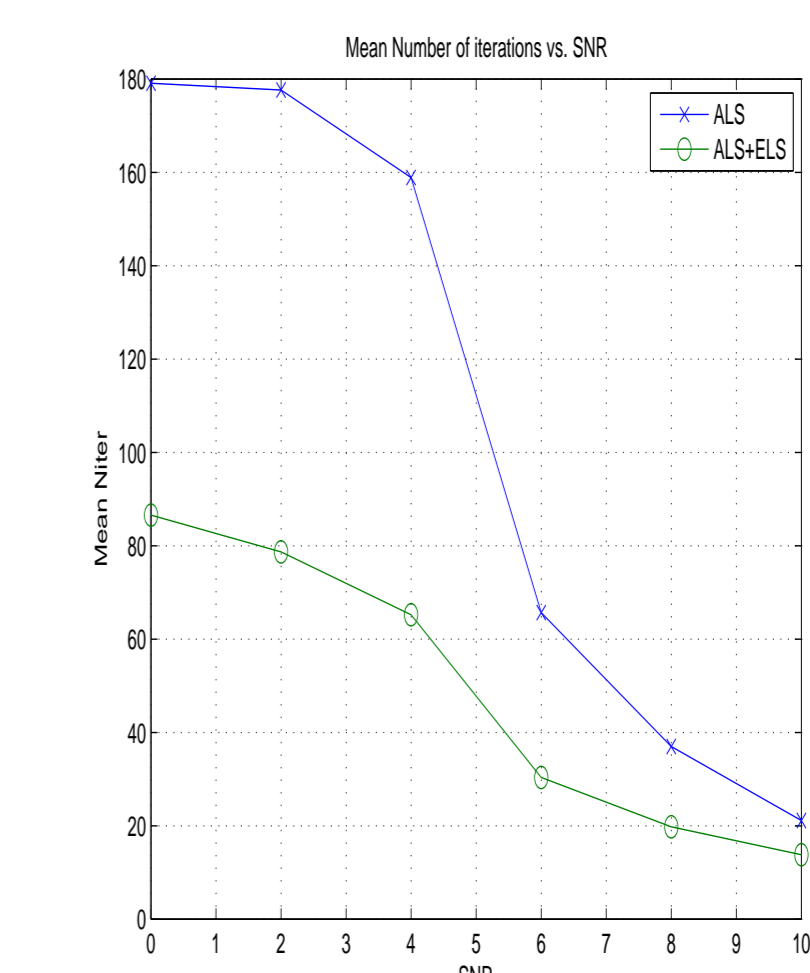
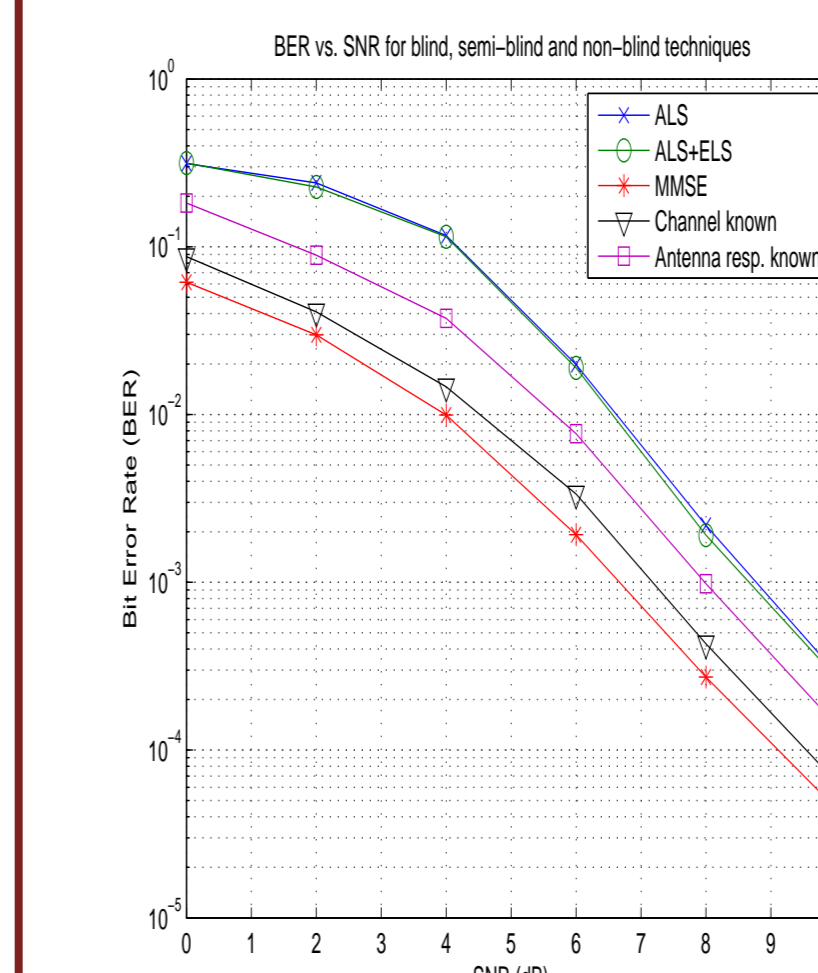
This scheme is inserted in the standard ALS algorithm scheme as follows:

**Summary of the ALS + ILS algorithm:**

- 1- Initialize  $\mathbf{A}^{(n-2)}, \mathbf{S}^{(n-2)}, \mathbf{H}^{(n-2)}, \mathbf{A}^{(n-1)}, \mathbf{S}^{(n-1)}, \mathbf{H}^{(n-1)}, n = 2$ .
  - 2- ILS Scheme:
    - Find the optimal value of  $\rho$  from (i) and (ii).
    - Build  $\mathbf{A}^{(new)}, \mathbf{S}^{(new)}$  and  $\mathbf{H}^{(new)}$  from (3).
  - 3- ALS Steps:
    - Find  $\mathbf{A}^{(n)}$  from  $\mathbf{S}^{(new)}$  and  $\mathbf{H}^{(new)}$ .
    - Find  $\mathbf{S}^{(n)}$  from  $\mathbf{A}^{(n)}$  and  $\mathbf{H}^{(new)}$ .
    - Find  $\mathbf{H}^{(n)}$  from  $\mathbf{A}^{(n)}$  and  $\mathbf{S}^{(n)}$ .
  - 4- Repeat from 2 until  $c(n) < \epsilon$  (e.g.  $\epsilon = 10^{-5}$ ), where  $c(n) = \|\mathcal{Y}^{(n)} - \mathcal{Y}^{(n-1)}\|^2$ .
- Increase  $n$  to  $n + 1$

## Experimental Results

- Performance in presence of AWGN. Noisy tensor of observation:  $\mathcal{Y}_{obs} = \mathcal{Y} + \mathcal{N}$ .
- Parameters:  $I = K = 6, J = 30$  QPSK symbols,  $L = P = 2, R = 4$  (On the uniqueness bound).
- Comparison between performance of BFM Blind Receiver, MMSE (Non-Blind) Receiver, and Semi-Blind Receivers (either  $\mathbf{H}$  or  $\mathbf{A}$  known).



## Conclusion

The Block Factor Model is a powerful blind receiver for multi-user access in wireless communications, with performance close to the MMSE receiver. The Iterative Line Search scheme greatly improves the convergence speed of the standard ALS algorithm [1]. Another faster algorithm has been developed in [2].

[1] Dimitri Nion and Lieven De Lathauwer, "A Block Factor Analysis based Receiver for Blind Multi-User Access in Wireless Communications", ICASSP 2006, May, Toulouse, France.

[2] Dimitri Nion and Lieven De Lathauwer, "Levenberg-Marquardt Computation of the Block Factor Model for Blind Multi-User Access in Wireless Communications", EUSIPCO 2006, September 4-8, Florence, Italy, accepted.

[3] Myriam Rajih and Pierre Comon, "Enhanced Line Search: A novel Method to Accelerate PARAFAC", EUSIPCO 2005, September 4-8, Antalya, Turkey.